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OPTIMIZED MICROSTRIP RING-STAR 5-PORTS FOR BROADBAND 6-PORT MEASUREMENT APPLICATIONS

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Summary

Results of computer-aided analysis and optimization of symmetrical 5-port microstrip rings with external matching and inner star coupling are presented. Over one octave operational bandwidth with power balance within 1dB and matching better than -20dB has been achieved in theory and measurement for two different design concepts.

Introduction

Well matched broadband couplers could be of great practical significance for 6- resp. 5-port measurement systems (1). stripline or microstrip techniques may be used to construct compact, cheap and simple 5-ports up to the K_u -band frequency range. Superposition of high impedance star- (fig. 1) and ring-5-ports (fig. 2) may be used (2,3) to achieve over one octave bandwidth. Dielectrical separation (Mylar-spacer) of the two coupler components has been used (2), producing capacitive resonance coupling of the ring- and the star-element, thus broadening the device bandwidth. However, most significant design parameters for broadband operation are the external matching circuits at the ports, together with 'inner' coupling (matching) of the star component to the ring.

Fig.1 star 5-port

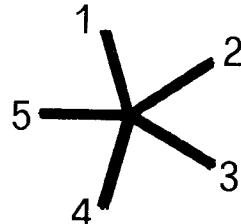
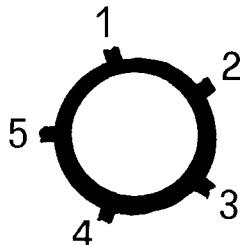


Fig.2 ring 5-port



The use of one of the matching concepts alone reveals much lower intrinsic bandwidth (4) under the condition of 20dB return loss. We have made a complete circuit analysis and optimization of ring-star 5-port components including specifically the effect of 'inner' coupling between star and ring combined with additional external matching by line segments and lumped capacitances.

The 5-port coupler design

Lossless 5-ports with a unitarian scattering matrix according to the condition:

$$S^{\#t} \cdot S = E \quad (1)$$

intrinsically feature equal power split $|S_{ik}| = 0.5$ with a phase relation of $\pm 120^\circ$ between port 2 and 3 or port 4 and 5 when symmetry, reciprocity and ideal matching at all ports is assured. The 'q-points' (1) for such a 5-port then are ideally situated in the complex power detection plane for 6-port measurements. For simple ring-type 5-ports this is achieved only at a single center-frequency ($\beta_1 = 75, 53^\circ$) and with a characteristic ring-impedance of $Z = 44, 72 \Omega$, when $Z_e = 50 \Omega$ microstrip-line system is presumed for the coupler ports. An ideal star-type coupler provides balanced powersplit at all frequencies, but does not fulfill the 120° phase condition, hence cannot be matched at all ports simultaneously. However, significant improvement of bandwidth is obtained by a combination of these ring- and star-type couplers. We have optimized this 5-port device using computer aided design methods. Calculations have been performed on the basis of standard microstrip transmission-line theory including dispersion of the effective dielectric constant in microstrip. The interaction of ring and star, mainly determined by the value of coupling C 's, is simplified as following: Optimum matching will be obtained with dominant ring operation ($C_g \rightarrow 0$ with $Z_i \gg Z$), but going along with small matching bandwidth known

from isolated ring structures. On the other hand the superposition of the star effects ($60 \text{ Ohms} > X_C = 1/\omega C_g > 27 \text{ Ohms}$) will lead to a broadband power balance and return loss niveau, albeit at the expense of the ideal matched condition.

A microstrip ring-star device has been developed (fig. 3) with the inner star rotation angle and coupling capacitances C_g , C_g' as additional parameters for coupler design. Best performance has been achieved with symmetrically centered star position (36°) between the outer ports. Optimum broadband operation of the device in fig. 3 has been found with the high-impedance star ($Z_i = 114 \text{ Ohms}$) coupled to the ring ($Z = 69 \text{ Ohms}$) by lumped capacitances $C_g' = 1 \text{ pF}$ for 'internal' matching purposes. The stubs ($Z_s = Z_i$) are used for fine-tuning the electrical length ratio β_1/β_{1i} of ring and star.

The outgoing lines ($Z_e = 50 \text{ Ohms}$) can be broadband matched to the ring-star device by short line-segments ($Z_h = 114 \text{ Ohms}$) and lumped capacitances $C_g \leq 1 \text{ pF}$. Calculations result in over one octave bandwidth (fig. 4) for the port transmission S_{ik} . Input reflection S_{ii} under -20 dB is observed in a range of 4-8 GHz. Fig. 5 shows good power balance of the ports ($S_{12}:S_{13} < 2 \text{ dB}$) and extremely flat characteristic of the port-phase-condition for exactly one octave (4-8 GHz), proving the broadband matched condition.

Our optimization analysis has revealed the following design criteria for ring/star-type couplers:

-center frequency is determined by the choice of the ring-segment length l . Best microstrip coupler performance is observed for $\beta_1/\beta_{1i} = 1,4$ with $\beta l = 110^\circ$, $\beta_{1i} = 80^\circ$, which requires a slightly smaller l_i compared to simple geometric consideration: $l/l_i = 2\pi/5 = 1,256$. For optimum matched condition the relation of the characteristic impedances Z/Z_i has to be chosen in such a way that a parallel connection of the two results in the characteristic impedance $Z = 44,72 \text{ Ohms}$ of the ideal ring coupler. Additionally the highest Z_i forms the best design with the restriction $Z_i < 120 \text{ Ohms}$, due to microstrip technological constraints.

The coupling elements C_g, C_g' play a key role for broadband, low reflection operation, since for low frequencies the 120° phase relation requires a rather compact reactance. The coupling C's should be in the order of $0,5-1,5 \text{ pF}$, $X_C = 0,4 \cdot Z$ at the low frequency end.

In addition to these 'inner' coupling effects, 'external' matching is necessary for smooth broadband performance: The short high impedance line-segments $Z_h \approx Z_i$ with optimum $\beta L = 7,4^\circ$ are used for fine-tuning the coupler matching characteristics.

A modified 5-port design is shown in fig. 6, where the inner coupling capacitances are replaced by linear taper sections, thus simplifying the 5-port construction and eliminating the parasitics and tolerances of lumped elements. Matching bandwidth, power balance and phase condition is found excellent also over one octave, fig. 7,8. In contrast to the first design the resonance at the low frequency end of the operational band has been removed due to broadband matching instead of resonance matching the star and the ring.

Conclusion

Microstrip 5-port ring-star-type devices can be designed to operate as broadband detection devices for use in 6-port measurement systems. An analysis of the influences of all parameters has clarified the operational principle of these 5-port-devices. Computer aided optimization of 5-port parameters result in two design concepts that reveal one octave operational bandwidth in theory and experiments. These elements are ideally suited for the integration of detector diodes, then forming a compact detection device for small and cheap 6-port measurement systems.

References

- (1) G.F. Engen, 'The six-port reflectometer: An alternative network analyzer', IEEE Trans. MTT, vol. MTT-25, pp. 1075 - 1080
- (2) F.C. de Ronde, 'Octave-wide matched symmetrical, reciprocal 4- and 5-ports', MTT-82 Symp., June 1982, Dallas
- (3) M. Malkomes, H.J. Schmitt, G. Padisch, M. van der Poel, 'Broadband Multi-port-Coupler', 13th European Microwave Conf., Nürnberg, pp. 339 - 343
- (4) E.R.B. Hansson, G.P. Riblet, 'An Ideal Six-Port Network Consisting of a Matched Reciprocal Lossless Five-Port and a Perfect Directional Coupler', IEEE Trans. MTT, vol. MTT-31, pp. 284 - 288, March 1983

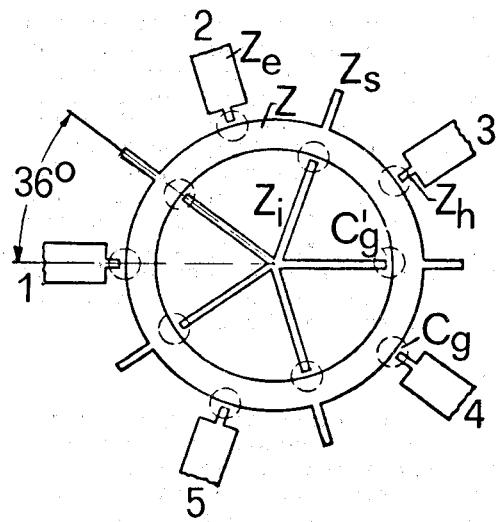


Fig.3 Microstrip ring-star 5-port

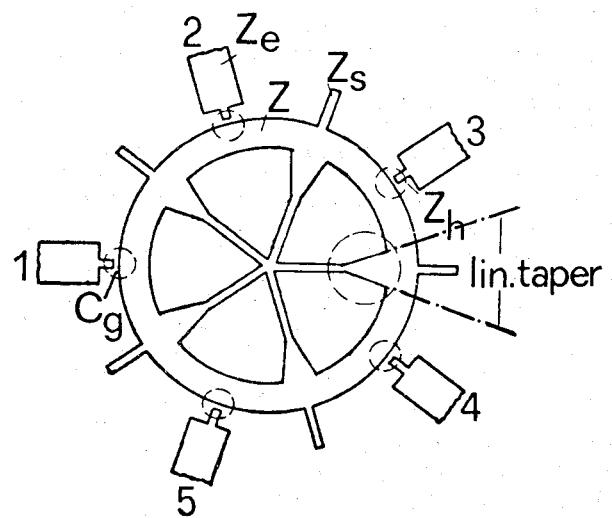


Fig.6 Microstrip ring-star 5-port

Fig.4 Scattering parameters of 5-port in fig.3

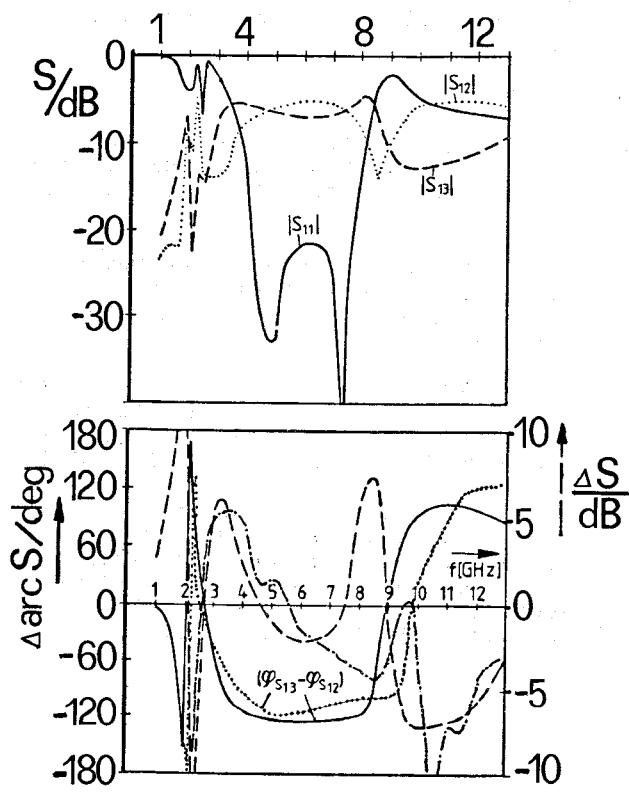


Fig.5 Power balance and phase condition of 5-port in fig.3

Theory ————
Exp. ————

Fig.7 Scattering parameters of 5-port in fig.6

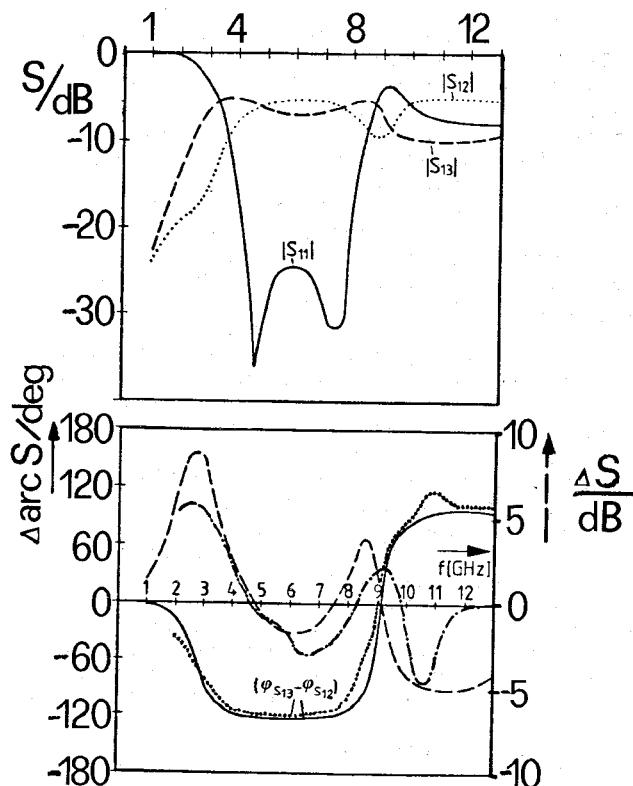


Fig.8 Power balance and phase condition of 5-port in fig.6

Theory ————
Exp. ————